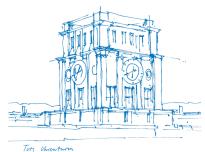


## A Bilinear Equalizer for Massive MIMO Systems

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1see: arXiv:1707.09940



# 1. Bilinear Equalization



## **CSI** Acquisition

transmission of orthogonal pilots estimation of channel by correlation:

$$oldsymbol{arphi}_k = oldsymbol{h}_k + rac{1}{\sqrt{
ho_{\mathsf{tr}}}} oldsymbol{w}_k$$



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MMSE estimate:

$$\hat{m{h}}_k = m{C}_k m{Q}_k^{-1} m{arphi}_k$$

with 
$$m{C}_k = \mathrm{E}[m{h}_k m{h}_k^{\mathrm{H}}]$$
 and  $m{Q}_k = m{C}_k + rac{1}{
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#### **Pilot Contamination**

transmission of orthogonal pilots but more users than pilots reuse of pilots

interference in estimation of channel by correlation:

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## Data Transmission and Equalization

received signal in the uplink

$$oldsymbol{y} = \sum_k \sqrt{p}_k oldsymbol{h}_k s_k + oldsymbol{v}$$

with  $s_k \sim \mathcal{N}_{\mathbb{C}}(0,1)$  and  $\boldsymbol{v} \sim \mathcal{N}_{\mathbb{C}}(\mathbf{0},\mathbf{I})$ 



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data estimate for user k:

$$egin{aligned} \hat{s}_k &= oldsymbol{g}_k^{\mathrm{H}} oldsymbol{y} \ &= \sqrt{p_k} oldsymbol{g}_k^{\mathrm{H}} oldsymbol{h}_k s_k + \sum_{j 
eq k} \sqrt{p_j} oldsymbol{g}_k^{\mathrm{H}} oldsymbol{h}_j s_j + oldsymbol{g}_k^{\mathrm{H}} oldsymbol{v} \end{aligned}$$



#### Equalization with Matched Filter

estimate based on matched filter (good equalization for massive MIMO)

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use channel estimate of correlator

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estimate based on matched filter (good equalization for massive MIMO)

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or MMSE channel estimate

$$\hat{s}_k = oldsymbol{arphi}_k^{\mathrm{H}} oldsymbol{Q}_k^{-1} oldsymbol{C}_k oldsymbol{y}$$



### Bilinear Equalization

depending on the employed channel estimator

$$oldsymbol{g}_k^{ ext{H}} = oldsymbol{arphi}_k^{ ext{H}} oldsymbol{A}_k$$

with deterministic  $A_k$  (independent of actual CSI)



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with deterministic  $\boldsymbol{A}_k$  (independent of actual CSI) therefore, generalized matched filter



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$$oldsymbol{g}_k^{ ext{H}} = oldsymbol{arphi}_k^{ ext{H}} oldsymbol{A}_k$$

with deterministic  $A_k$  (independent of actual CSI)

estimate

$$\hat{s}_k = oldsymbol{arphi}_k^{ ext{H}} oldsymbol{A}_k oldsymbol{y}$$

is linear in  $oldsymbol{y}$  and also linear in  $oldsymbol{arphi}_k$ 

thus, bilinear equalization



# 2. Asymptotic Properties of SINR Maximization



use Medard's bound

$$I(s_k; \hat{s}_k) \ge \log_2(1 + \gamma_k^{\mathsf{ul}})$$

with the SINR  $\gamma_k^{\rm ul}$ 



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optimal SINR can be written as

$$\gamma_k^{\star} = M p_k \frac{\boldsymbol{e}_k^{\mathrm{T}} \boldsymbol{\Gamma}(\frac{1}{M} \boldsymbol{P}^{-1} + \boldsymbol{\Gamma})^{-1} \boldsymbol{e}_k}{\boldsymbol{e}_k^{\mathrm{T}}(\frac{1}{M} \boldsymbol{P}^{-1} + \boldsymbol{\Gamma})^{-1} \boldsymbol{e}_k}$$

with  $[\Gamma]_{nk}=rac{1}{M} \; {
m tr} \left(m{C}_nm{Z}^{-1}m{C}_km{Q}_k^{-1}
ight)$  and  $m{Z}=m{I}+\sum_n p_nm{C}_n$ 



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for large M,

$$\gamma_k^{\star} o M \, rac{p_k}{oldsymbol{e}_k^{\mathrm{T}} oldsymbol{\Gamma}^{-1} oldsymbol{e}_k} = \gamma_k^{\mathrm{asy}}$$



## **Asymptotic Properties**

for linearly independent covariance matrices  $C_k$ ,  $k\in\Omega_p$  and  $\limsup_{M\to\infty}\|C_k\|_2<\infty$ 



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it can be shown that

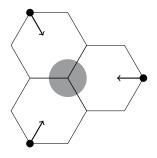
$$\liminf_{M \to \infty} \frac{\gamma_k^{\mathrm{asy}}}{M} > 0$$

has been demonstrated for the SINR depending on CSI in [Björnson, Hoydis, Sanguinetti, arXiv:170500538, May 2017]

also holds for application of approximate covariance matrices  $C_k$ 



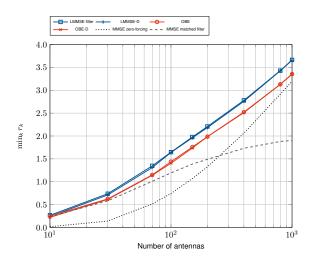
#### Simulation Scenario



5 users in each cell, only 5 pilots



#### Simulation w.r.t. Number of BS Antennas





#### Simulation w.r.t. SNR

